

and an input pulse (λ_p). When the flip-flop is in “state 0”, there is no FWM in the SOA, and the input pulse can not pass the AND gate. Variable attenuators (VOA) and a polarization controller (PC) are used to adjust the input power and the polarization state respectively, to maximize the FWM efficiency. The polarization dependence of the AND gate could be eliminated by using polarization diversity technique.

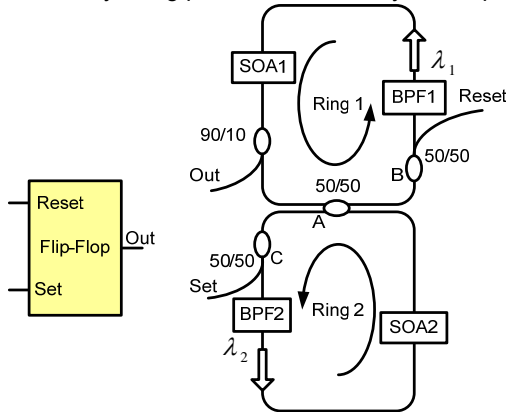


Fig. 3: All-optical flip-flop based on SOA fiber ring lasers

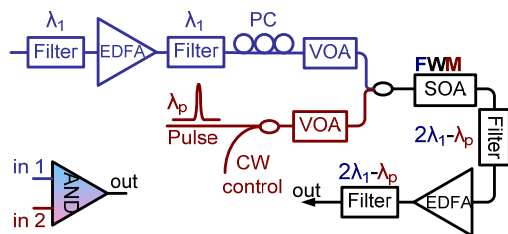


Fig. 4: All-optical AND logic gate based on FWM in SOA

Experimental results

The input clock pulse has a wavelength of $\lambda_p=1554.1$ nm, and the output of flip-flop 1 is $\lambda_1=1552.5$ nm. After FWM in “AND 1”, the carry 1 signal has a wavelength of $\lambda_{c1}=2\lambda_1-\lambda_p=1550.9$ nm. The output of flip-flop 2 is $\lambda_2=1549.3$ nm, and after FWM in “AND 2”, the carry 2 signal is $\lambda_{c2}=2\lambda_2-\lambda_{c1}=1547.7$ nm. All the filters we used have a 3-dB bandwidth of 0.8 nm.

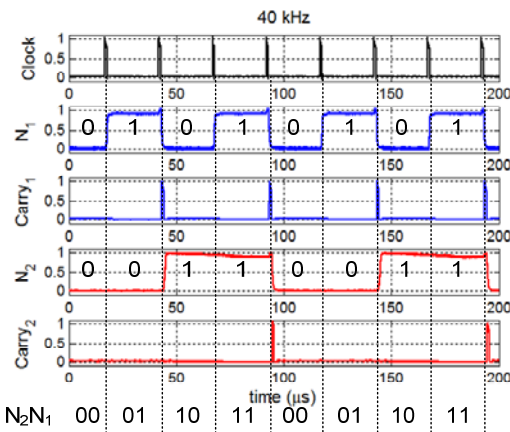


Fig. 5: Outputs of all-optical counter

In Fig.5, all-optical pulse counting is demonstrated. The input pulses have a repetition rate of 40 kHz with 1- μ s pulse-width. When a pulse inputs, N_2N_1 adds 1, from 00 to 01, then to 10, 11, and finally back to 00,

having a good agreement with Fig.2. In Fig.5, N_1 is a square waveform with a duty-cycle of 50% and a repetition rate of 20 kHz; N_2 is also a 50% square waveform but with a repetition rate of 10 kHz. Both carry 1 and carry 2 have the same pulse-width as the input pulse, but with repetition rates of 20 kHz and 10 kHz respectively, only 1/2 and 1/4 with respect to the input. Therefore, this scheme can also be used as an optical frequency divider.

Compared with the input pulse, it can be anticipated that both carry signals will have a CW pedestal due to the ASE noise of SOA in AND logic gates. In order to reduce these pedestals, in Fig.4, a CW light is used to saturate the SOA. It is confirmed in Fig.5 that there are no significant CW pedestals in both carry signals. In addition, Q-factor measurement is also carried out to evaluate the signal degeneration in each stage and investigate the stage cascading. The Q-factors of N_1 and N_2 are 16.1 and 19.9 respectively, only depending on the properties of flip-flops. The Q-factor of input pulse is 32.5. Due to the low repetition rate and gain-tilt mechanism of EDFA, carry 1 has a Q-factor of 17.0. By exploiting the aforementioned CW light to saturate the SOA, we are able to obtain carry 2 with a Q-factor of 15.0, only slightly lower than carry 1, further confirming the cascading of this scheme. The pulse counting rate of this scheme is mainly limited by the state switching time of optical flip-flops, which is due to the discrete device implementation of flip-flops. In our case, the cavity length of fiber lasers in flip-flops is 40 m, corresponding to ~ 1 μ s switching time. Photonic integration can reduce the switching time lower than 100 ps by shortening the cavity length to millimetres and make GHz counting possible⁵.

Conclusions

For the first time as we know, an all-optical pulse counter is demonstrated. By cascading identical stages composed by a SOA fiber laser flip-flop and an optical AND logic gate, we present an optical counter, which can also be used as an optical frequency divider. Two-bit optical pulse counting is implemented, as well as 1/2 and 1/4 optical frequency division. Q-factor measurements confirm the feasibility of the proposed solution and the cascading of single stage in order to increase the maximum bit-number of counting. Finally, photonic integration of optical flip-flops would allow the counting speed beyond GHz.

References

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